# A Qualitative Approach to Inverse Scattering for Anisotropic Media 

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## Scattering by an Inhomogeneous Media



The matrix valued function $A$ with $C^{1}(D)$ entries and $n \in L^{\infty}(D)$ are such that $\Re(A) \geq \alpha>0, \Im(A) \leq 0, \Re(n)>0$ and $\Im(n) \geq 0$. Here $k$ is the wave number and is proportional to the frequency $\omega, u^{i}$ is the incident wave and $S$ is the unit sphere.

$$
\nabla \cdot A \nabla u^{s}+k^{2} n u^{s}=\nabla \cdot(I-A) \nabla u^{i}+k^{2}(1-n) u^{i} \quad \text { in } \mathbb{R}^{3} .
$$

$A, n$ are extended by $I, 1$ respectively and $u^{s}:=u-u^{i}$ in $D$.

## Far Field Operator

## Scattering Data

$u_{\infty}(\hat{x}, d, k)$, for $d \in S_{i} \subset S, \hat{x} \in S_{m} \subset S$ and (possibly) $k \in\left[k_{1}, k_{2}\right]$.

The far field operator $F: L^{2}(S) \rightarrow L^{2}(S)$ is defined by

$$
(F g)(\hat{x}):=\int_{S} u_{\infty}(\hat{x}, d, k) g(d) d s_{d}
$$

■ Fg is the far field pattern of the scattered field corresponding to the incident field

$$
v_{g}(x):=\int_{S} e^{i k x \cdot d} g(d) d s_{d}
$$

(known as a Herglotz wave function).
■ $F$ is related to the scattering operator $\mathcal{S}$ by

$$
\mathcal{S}=I+\frac{i k}{2 \pi} F
$$

## Far Field Operator

## Theorem

$F: L^{2}(S) \rightarrow L^{2}(S)$ is injective and has dense range if and only if there does not exist a nontrivial solution to the transmission eigenvalue problem

$$
\begin{array}{cll}
\Delta v+k^{2} v=0 & \text { in } & D \\
\nabla \cdot A \nabla w+k^{2} n w=0 & \text { in } & D \\
w=v & \text { on } & \partial D \\
\nu \cdot A \nabla w=\nu \cdot \nabla v & \text { on } & \partial D
\end{array}
$$

such that $v:=v_{g}$ is a Herglotz wave function.

Values of $k \in \mathbb{C}$ for which the transmission eigenvalue problem has non trivial solution are called transmission eigenvalues.

Transmission eigenvalues relates to non-scattering frequencies.

## Qualitative Methods for the Support

- The linear sampling method has been widely used for various inverse scattering problems, limited aperture data etc.
© CAKOni-Colton (2014), A Qualitative Approach to Inverse Scattering Theory, Springer.
- Factorization methods is mathematically rigorous for exact data and justified for noisy data.

A A. Kirsch and N. Grinberg (2008), The Factorization Method for Inverse Problems, Oxford University Press.

- The generalized linear sampling method.

目 Audibert - Haddar (2014) - Inverse Problems.

All these method explore the (linear) far field operator to construct an indicator function for the support $D$ of the inhomogeneity

[^0]
## Transmission Eigenvalue Problem

Having determined the support $D$ without knowing anything about the material properties we would like to get some information about the constitutive parameters $A$ and $n$.

For this we appeal to the transmission eigenvalue problem for $v \in H^{1}(D)$ and $w \in H^{1}(D)$ such that

$$
\begin{array}{clc}
\Delta v+k^{2} v=0 & \text { in } & D \\
\nabla \cdot A \nabla w+k^{2} n w=0 & \text { in } & D \\
w=v & \text { on } & \partial D \\
\nu \cdot A \nabla w=\nu \cdot \nabla v & \text { on } & \partial D
\end{array}
$$

## Related Questions

- Connect transmission eigenvalues to $A$ and $n$.

■ Determine transmission eigenvalues from scattering data?

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Analysis of the Interior Transmission Problem and the Transmission Eigenvalue Problem

## Interior Transmission Problem

From now on to introduce our ideas we assume that both $A$ and $n$ are real valued.

The interior transmission problem reads: Find $v \in H^{1}(D)$ and $w \in H^{1}(D)$ such that

$$
\begin{array}{cll}
\Delta v+k^{2} v=\ell_{1} & \text { in } & D \\
\nabla \cdot A \nabla w+k^{2} n w=\ell_{2} & \text { in } & D \\
w-v=0 & \text { on } & \partial D \\
\nu \cdot A \nabla w-\nu \cdot \nabla v=h & \text { on } & \partial D
\end{array}
$$

for $\ell_{1} \in L^{2}(D), \ell_{2} \in L^{2}(D)$ and $h \in H^{-1 / 2}(\partial D)$.

## Notations

$$
\begin{aligned}
a_{\min }:= & \inf _{x \in D|\xi|=1} \inf \xi \cdot A(x) \xi>0, \\
& \text { and } \quad a_{\max }:=\sup _{x \in D} \sup _{|\xi|=1} \xi \cdot A(x) \xi<\infty . \\
& n_{\min }:=\inf _{x \in D} n(x)>0 \quad \text { and } \quad n_{\min }:=\sup _{x \in D} n(x)<\infty .
\end{aligned}
$$

Consider a $\delta$-neighborhood $\mathcal{N}$ of the boundary $\partial D$

$$
\mathcal{N}:=\{x \in D: \operatorname{dist}(x, \partial D)<\delta\}
$$

$$
\begin{aligned}
& a_{\star}:=\inf _{x \in \mathcal{N}|\xi|=1} \inf \xi \cdot A(x) \xi>0 \quad \text { and } \quad a^{\star}:=\sup _{x \in \mathcal{N}} \sup _{|\xi|=1} \xi \cdot A(x) \xi<\infty \\
& n_{\star}:=\inf _{x \in \mathcal{N}} n(x)>0 \text { and } n^{\star}:=\sup _{x \in \mathcal{N}} n(x)<\infty .
\end{aligned}
$$

## Modified Interior Transmission Problem

The modified transmission eigenvalue problem

$$
\begin{array}{clc}
\Delta v-\kappa^{2} v=\ell_{1} & \text { in } & D \\
\nabla \cdot A \nabla w-\kappa^{2} n_{0} w=\ell_{2} & \text { in } & D \\
w-v=0 & \text { on } & \partial D \\
\nu \cdot A \nabla w-\nu \cdot \nabla v=h & \text { on } & \partial D
\end{array}
$$

for some choice of $\kappa>0$ and $n_{0}>0$ is a compact perturbation of the interior transmission problem in

$$
\mathbf{H}(D):=\left\{(w, v) \in H^{1}(D) \times H^{1}(D): w-v \in H_{0}^{1}(D)\right\} .
$$

In variational form

$$
\begin{aligned}
& \int_{D} A \nabla w \cdot \nabla \bar{w}^{\prime} d x-\int_{D} \nabla v \cdot \nabla \bar{v}^{\prime} d x+\kappa^{2} \int_{D} n_{0} w \bar{w}^{\prime} d x-\kappa^{2} \int_{D} v \bar{v}^{\prime} d x \\
& =\int_{\partial D} h \overline{w^{\prime}} d s+\int_{D} \ell_{1} \overline{v^{\prime}} d x-\int_{D} \ell_{2} \overline{w^{\prime}} d x, \quad \text { for all } \quad\left(w^{\prime}, v^{\prime}\right) \in \mathbf{H}(D) .
\end{aligned}
$$

## Modified Interior Transmission Problem

Assume that either $a^{\star}<1$ and choose $n_{0}<1$, or $a_{\star}>1$ and choose $n_{0}>1$. Then for $\kappa>0$ large enough the sesquilinear form

$$
\begin{aligned}
& a\left((w, v),\left(w^{\prime}, v^{\prime}\right)\right):= \\
& \int_{D} A \nabla w \cdot \nabla \bar{w}^{\prime} d x-\int_{D} \nabla v \cdot \nabla \bar{v}^{\prime} d x+\kappa^{2} \int_{D} n_{0} w \bar{w}^{\prime} d x-\kappa^{2} \int_{D} v \bar{v}^{\prime} d x
\end{aligned}
$$

is $T$-coercive, i.e. $a^{T}\left((w, v),\left(w^{\prime}, v^{\prime}\right)\right):=a\left((w, v), \mathbf{T}\left(w^{\prime}, v^{\prime}\right)\right)$ is coercive with the isomorphism $\mathbf{T}: \mathbf{H}(D) \rightarrow \mathbf{H}(D)$ defined by

$$
\mathbf{T}:(w, v) \mapsto(w-2 \chi v,-v) \quad \text { or } \quad \mathbf{T}:(w, v) \mapsto(w,-v+2 \chi w),
$$

respectively, where $\chi$ is $C^{\infty}$ cut off function supported in $\overline{\mathcal{N}}$.

## Proof on the board

目 Bonnet-Ben Dhia - Chesnel, Lucas - Haddar (2011) - C. R. Math. Acad. Sci. Paris

## Transmission Eigenvalue Problem

If either $a^{\star}<1$ or $a_{\star}>1$ then the interior transmission problem is well posed provided that $k \in \mathbb{C}$ is not a transmission eigenvalue.

Under the above assumptions, to show discreteness of transmission eigenvalues it suffices to find one $k \in \mathbb{C}$ that is not a transmission eigenvalue.

If either $a^{\star}<1$ and $n^{\star}<1$, or $a_{\star}>1$ and $n_{\star}>1$ then the set of transmission eigenvalues is discrete in $\mathbb{C}$ with $+\infty$ as the only possible accumulation point.

If either $a_{\max }<1$ or $a_{\text {min }}>1$, and $\int_{D}(n-1) d x \neq 0$, then the set of transmission eigenvalues is discrete in $\mathbb{C}$ with $+\infty$ as the only possible accumulation point.

## Transmission Eigenvalue Problem: $n \equiv 1$ case.

The transmission eigenvalue problem for $n \equiv 1$ can be written for $\mathbf{w}=A \nabla w \in L^{2}(D), \mathbf{v}=\nabla v \in L^{2}(D)$ and $N:=A^{-1}$ as

$$
\begin{array}{clc}
\nabla(\nabla \cdot \mathbf{v})+k^{2} \mathbf{v}=0 & \text { in } & D \\
\nabla(\nabla \cdot \mathbf{w})+k^{2} N \mathbf{w}=0 & \text { in } & D \\
\nu \cdot \mathbf{w}=\nu \cdot \mathbf{v} & \text { on } & \partial D \\
\nabla \cdot \mathbf{w}=\nabla \cdot \mathbf{v} & \text { on } & \partial D
\end{array}
$$

with $\mathbf{w}-\mathbf{v} \in \mathcal{H}_{0}(D)$ where

$$
\begin{aligned}
H_{0}(\operatorname{div}, D): & =\left\{\mathbf{u} \in L^{2}(D)^{2}, \nabla \cdot \mathbf{u} \in L^{2}(D), \nu \cdot \mathbf{u}=0 \text { on } \partial D\right\} \\
\mathcal{H}_{0}(D): & =\left\{\mathbf{u} \in H_{0}(\operatorname{div}, D): \nabla \cdot \mathbf{u} \in H_{0}^{1}(D)\right\} .
\end{aligned}
$$

which for $\mathbf{u}:=\mathbf{w}-\mathbf{v} \in \mathcal{H}_{0}(D)$ is equivalent to
$\int_{D}(N-I)^{-1}\left(\nabla \nabla \cdot \mathbf{u}+k^{2} \mathbf{u}\right) \cdot\left(\nabla \nabla \cdot \overline{\mathbf{u}^{\prime}}+k^{2} N \overline{\mathbf{u}^{\prime}}\right) d x=0, \quad \forall \mathbf{u}^{\prime} \in \mathcal{H}_{0}(D)$.

## Transmission Eigenvalue Problem

At this point we assume that either $a_{\max }<1$ or $a_{\min }>1$. and consider only $k>0$.

Take $a_{\max }<1$ which implies that $\xi \cdot(N-I)^{-1} \xi \geq \alpha|\xi|^{2}, \alpha=\frac{a_{\max }}{1-a_{\max }}$.

$$
\mathcal{A}_{k}\left(\mathbf{u}, \mathbf{u}^{\prime}\right):=\left((N-I)^{-1}\left(\nabla \nabla \cdot \mathbf{u}+k^{2} \mathbf{u}\right),\left(\nabla \nabla \cdot \mathbf{u}^{\prime}+k^{2} \mathbf{u}^{\prime}\right)\right)_{D}+k^{4}\left(\mathbf{u}, \mathbf{u}^{\prime}\right)_{D}
$$

$$
\mathcal{B}\left(\mathbf{u}, \mathbf{u}^{\prime}\right):=\left(\nabla \cdot \mathbf{u}, \nabla \cdot \mathbf{u}^{\prime}\right)_{D} .
$$

Here $(\cdot, \cdot)_{D}$ denotes the $L^{2}(D)$-inner product.
The eigenvalue problem becomes

$$
\mathcal{A}_{k}\left(\mathbf{u}, \mathbf{u}^{\prime}\right)-k^{2} \mathcal{B}\left(\mathbf{u}, \mathbf{u}^{\prime}\right)=0 \quad \text { or } \quad \mathbb{A}_{k} \mathbf{u}-k^{2} \mathbb{B} \mathbf{u}=0
$$

$\left(\mathbb{A}_{k} \mathbf{u}, \mathbf{u}^{\prime}\right)_{\mathcal{H}_{0}(D)}=\mathcal{A}_{k}\left(\mathbf{u}, \mathbf{u}^{\prime}\right) \quad$ and $\quad\left(\mathbb{B} \mathbf{u}, \mathbf{u}^{\prime}\right)_{\mathcal{H}_{0}(D)}=\mathcal{B}\left(\mathbf{u}, \mathbf{u}^{\prime}\right)$.

## Transmission Eigenvalue Problem

$$
\begin{aligned}
\mathcal{A}_{k}(\mathbf{u}, \mathbf{u})-k^{2} \mathcal{B}(\mathbf{u}, \mathbf{u}) & \geq\left(\alpha-\frac{\alpha^{2}}{\epsilon}\right)\|\nabla \nabla \cdot \mathbf{u}\|_{L^{2}(D)}^{2}+(1+\alpha-\epsilon) k^{2}\|\mathbf{u}\|_{L^{2}(D)}^{2} \\
& -k^{2} \frac{1}{\lambda_{1}(D)}\|\nabla \nabla \cdot \mathbf{u}\|_{L^{2}(D)}^{2}
\end{aligned}
$$

hence from the Poincaré inequality

$$
\|\nabla \cdot \mathbf{u}\|_{L^{2}(D)}^{2} \leq \frac{1}{\lambda_{1}(D)}\|\nabla \nabla \cdot \mathbf{u}\|_{L^{2}(D)}^{2}
$$

there are no transmission eigenvalues if $k^{2}<\alpha /(1+\alpha) \lambda_{1}(D)$ where $\lambda_{1}(D)$ is the first Dirichlet eigenvalue of $-\Delta$ on $D$.

A Faber-Krahn type inequality for TE - All transmission eigenvalues satisfy

$$
k^{2}>\lambda_{1}(D) a_{\max }
$$

## Existence of Real Transmission Eigenvalues

■ The mapping $k \rightarrow \mathbb{A}_{k}$ is continuous from $(0,+\infty)$ to the set of self-adjoint coercive operators from $\mathcal{H}_{0}(D) \rightarrow \mathcal{H}_{0}(D)$.
$\square \mathbb{B}: \mathcal{H}_{0}(D) \rightarrow \mathcal{H}_{0}(D)$ is self-adjoint, compact and non-negative.
There exists an increasing sequence of eigenvalues $\lambda_{j}(k)_{j \geq 1}$ of the generalized eigenvalue problem

$$
\mathbb{A}_{k} u-\lambda(k) \mathbb{B} u=0 \quad \text { in } \mathcal{H}_{0}(D)
$$

such that

$$
\lambda_{j}(k)=\min _{W \subset \mathcal{U}_{j}} \max _{u \in W \backslash\{0\}} \frac{\left(\mathbb{A}_{k} u, u\right)}{(\mathbb{B} u, u)}
$$

where $\mathcal{U}_{j}$ denotes the set of all $j$-dimensional subspaces $W$ of $\mathcal{H}_{0}(D)$, $W \cap N(\mathbb{B})=\{0\}$

Then $k$ is a transmission eigenvalue if and only if satisfies

$$
\lambda_{j}(k)=k^{2}
$$

## Existence of Real Transmission Eigenvalues

Max-min principle for $\lambda_{j}(\tau)$ implies that if there exists $k_{0}>0$ and $k_{1}>0$ such that

■ $\mathbb{A}_{k_{0}}-k_{0}^{2} \mathbb{B}$ is positive on $\mathcal{H}_{0}(D)$,

- $\mathbb{A}_{k_{1}}-k_{1}^{2} \mathbb{B}$ is non positive on a $m$ dimensional subspace of $\mathcal{H}_{0}(D)$ then each $\lambda_{j}(k)=k^{2}$ for $j=1, \ldots, m$, has at least one solution in [ $k_{0}, k_{1}$ ], i.e. there exists $m$ transmission eigenvalues counting multiplicity within the interval $\left[k_{0}, k_{1}\right]$.

It is now obvious that determining such constants $k_{0}$ and $k_{1}$ provides the existence of transmission eigenvalues as well as the desired isoperimetric inequalities.

## Existence of Real Transmission Eigenvalues

## Theorem (CAKONI-GINTIDES-HADDAR)

Assume that $a_{\max }<1$. Then, there exists an infinite discrete set of real transmission eigenvalues $k_{j}$ accumulating at $+\infty$. Furthermore

$$
k_{j}\left(a_{\min }, B_{1}\right) \leq k_{j}\left(a_{\min }, D\right) \leq k_{j}(A(x), D) \leq k_{j}\left(a_{\max }, D\right) \leq k_{j}\left(a_{\min }, B_{2}\right)
$$

where $B_{2} \subset D \subset B_{1}$.

If $A:=a l, 1 \neq a>0$ is constant, the first transmission eigenvalue uniquely determines the constant index of refraction.

Similar results can be obtained for the case when $a_{\text {min }}>1$.

## Transmission Eigenvalues: $n \not \equiv 1$ case

The analysis of the existence of real transmission eigenvalues when $n \not \equiv 1$ is more complicated and restrictive.
國 Cakoni-Kırsch (2010) - Int. J. Comput. Sci. Math.

- If the contrasts $A-I$ and $n-1$ have the same fixed sign, then there exists an infinite discrete set of real transmission eigenvalues accumulating at $+\infty$.
- If the contrasts $A-I$ and $n-1$ have the opposite fixed sign, then there exits at least one real transmission eigenvalue providing that $n$ is small enough.

䍰 Harris-Cakoni-Sun (2014) - Inverse Problems
Assume that there is $D_{0} \subset D$ (void) where $A=I$ and $n=1$, otherwise in $A$ and $n$ satisfy the above assumption. Then there exists at least one real transmission eigenvalue provided that $D_{0}$ is sufficiently small and this eigenvalue is depends monotonically increasing on the void size.

## Transmission Eigenvalues: $n \not \equiv 1$ case

Set $u=w-v \in H_{0}^{1}(D)$. Find $v=v_{u}$ by solving a Neuman type problem: For every $\psi \in H^{1}(D)$

$$
\int_{D}(A-l) \nabla v \cdot \nabla \bar{\psi}-k^{2}(n-1) v \bar{\psi} d x=\int_{D} A \nabla u \cdot \nabla \bar{\psi}-k^{2} n u \bar{\psi} d x
$$

Having $u \rightarrow v_{u}$, we require that $v:=v_{u}$ satisfies $\Delta v+k^{2} v=0$.
Thus we define $\mathbb{L}_{k}: H_{0}^{1}(D) \rightarrow H_{0}^{1}(D)$

$$
\left(\mathbb{L}_{k} u, \phi\right)_{H_{0}^{1}(D)}=\int_{D} \nabla v_{u} \cdot \nabla \bar{\phi}-k^{2} v_{u} \cdot \bar{\phi} d x, \quad \phi \in H_{0}^{1}(D) .
$$

Then the transmission eigenvalue problem is equivalent to

$$
\begin{aligned}
& \mathbb{L}_{k} u=0 \quad \text { in } \quad H_{0}^{1}(D) \quad \text { which can be written } \\
& \quad\left(\mathbb{I}+\mathbb{L}_{0}^{-1 / 2} \mathbb{C}_{k} \mathbb{L}_{0}^{-1 / 2}\right) u=0 \quad \text { in } \quad H_{0}^{1}(D)
\end{aligned}
$$

$\mathbb{L}_{0}$ self-adjoint positive definite and $\mathbb{C}_{k}$ self-adjoint compact.

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## Determination of Transmission Eigenvalues

from Scattering Data

RuTGERS

## Determination of Transmission Eigenvalues

First approach is based on the Linear Sampling Method
R Cakoni-Colton-Haddar (2010) C. R. Math. Acad. Sci. Paris The linear sampling method explores the far field equation

$$
(F g)(\hat{x})=\Phi_{\infty}(\hat{x}, z, k), \quad \text { for } \quad g \in L^{2}(S), \quad z \in D, \quad k \in\left[k_{0}, k_{1}\right]
$$

As you know "solutions" to this equations are such that the Herglotz function $v_{z}:=v_{g}(x)=\int_{S} e^{i k x \cdot d} g(d) d s$ and $w_{z}$ solve

$$
\begin{array}{ccc}
\Delta v_{z}+k^{2} v_{z}=0 & \text { in } & D \\
\nabla \cdot A \nabla w_{z}+k^{2} n w_{z}=0 & \text { in } & D \\
w_{z}-v_{z}=\Phi(\cdot, z) & \text { on } & \partial D \\
\nu \cdot A \nabla w_{z}-\nu \cdot \nabla v_{z}=\nu \cdot \nabla \Phi(\cdot, z) & \text { on } & \partial D
\end{array}
$$

## Determination of Transmission Eigenvalues

$$
F g=B v_{g}
$$

with the compact operator $B:\left\{\Delta v+k^{2} v=0, v \in H^{1}(D)\right\} \rightarrow L^{2}(S)$

$$
B: u^{i} \mapsto u_{\infty}^{s}, \text { with } \nabla \cdot A \nabla u^{s}+k^{2} n u^{s}=\nabla \cdot(I-A) \nabla u^{i}+k^{2}(1-n) u^{i} .
$$

Hence we have that

$$
B v_{z}=\Phi_{\infty}(\hat{x}, z, k) .
$$

■ If $k$ is not a transmission eigenvalue there exists a sequence of $g_{\epsilon}^{z} \in L^{2}(D)$ such that

$$
\left\|F g_{\epsilon}^{z}-\Phi_{\infty}(\cdot, z, k)\right\|_{L^{2}(S)} \rightarrow 0 \quad \epsilon \rightarrow 0
$$

and the Herglotz function $v_{g_{\epsilon}^{2}} \rightarrow v_{z}$ in $H^{1}(D)$

- If $k$ is a transmission eigenvalue and $g_{\epsilon}^{z}$ as above, $v_{g_{\epsilon}^{z}}$ can not be bounded in $H^{1}(D)$ norm as $\epsilon \rightarrow 0$, for almost all $z \in D$.
Proof on the board


## Determination of Transmission Eigenvalues

Can the same be said about the Tikhonov regularized solution $g_{\delta}^{z}$ of the far field equation with noisy far field operator $F^{\delta}$, i.e. the unique minimizer $g_{\delta}^{z}$ of

$$
\left\|F^{\delta} g_{\delta}^{z}-\Phi_{\infty}(\cdot, z)\right\|_{L^{2}(S)}^{2}+\epsilon\left\|g_{\delta}^{z}\right\|_{L^{2}(S)}^{2}
$$

where $\epsilon$ is the Tikhonov regularization parameter?
If $F$ has dense range it is easy to show that

$$
\lim _{\delta \rightarrow 0}\left\|F^{\delta} g_{\delta}^{z}-\Phi(\cdot, z)\right\|_{L^{2}(S)}=0
$$

■ Thus for almost all $z \in D$, if $k$ is a transmission eigenvalue $\lim _{\delta \rightarrow 0}\left\|v_{g_{\delta}^{2}}\right\|_{H^{\prime}(D)}=\infty$.

■ If $k$ is not a transmission eigenvalue $\lim _{\delta \rightarrow 0}\left\|v_{g_{\delta}^{2}}\right\|_{H^{1}(D)}$ exists.
The proof of the latter involves the factorization method (Arens 2004)

## Computation of Transmission Eigenvalues


$D$ square $2 \times 2, A=I$ and $n=16$. The far field equation is solved for several source points $z$ inside $D$ using 42 incoming directions and measurements. Red dots indicate exact eigenvalues.

## Inside-Outside Duality

Characterize the transmission eigenvalues $k$ from the behavior of the eigenvalues of the far field operator $F_{k}: L^{2}(S) \rightarrow L^{2}(S)$

$$
\left(F_{k} g\right)(\hat{x}):=\int_{S} u_{\infty}(\hat{x}, d, k) g(d) d s_{d}
$$

囦 KIRSCh-Lechleiter (2013) - Inverse Problems
围 Lechleiter-Peters (2015) - Com. Math. Sci.
Essential is a symmetric factorization of the far field operator

$$
F_{k}=H_{k} \mathbf{T}_{k} H_{k}^{*}
$$

where (loosely) $H_{k}: L^{2}(S) \rightarrow \mathcal{X}_{k}(D)$ is such that $H_{k}^{*}$ has dense range, $\mathbf{T}_{k}: \mathcal{X}_{k}(D) \rightarrow \mathcal{X}_{k}(D)$ is compact perturbation of a coercive operator and its imaginary part satisfies a sign condition.

## Inside-Outside Duality

Assume that $A=I$ and either $n>1$ or $n<1$, or $n=1$ and either $A>I$ or $A<I$. We call $q$ the contrast, i.e. $q=n-1$ or $q=I-A$.
Facts on the compact operator $F_{k}$ (recall $\mathcal{S}_{k}=I+\frac{i k}{2 \pi} F_{k}$ ).
$\square$ For real $A$ and $n, F_{k}$ is normal, i.e. $F_{k} F_{k}^{*}=F_{k}^{*} F_{k}$. Thus, $\mathcal{S}_{k}$ is unitary, i.e. $\mathcal{S}_{k} \mathcal{S}_{k}^{*}=\mathcal{S}_{k}^{*} \mathcal{S}_{k}=I$.
■ As such $F_{k}$ has an infinite number of eigenvalues $\lambda_{j}(k)$ accumulating to 0 : they lie on the circle in $\mathbb{C}$

$$
|\lambda|^{2}-\frac{4 \pi}{k} \Im(\lambda)=0 .
$$

■ For $k$ not a transmission eigenvalue, as $j \rightarrow \infty$, $\lambda_{j}(k) /\left|\lambda_{j}(k)\right| \rightarrow-1$ if $q>0$ and $\lambda_{j}(k) /\left|\lambda_{j}(k)\right| \rightarrow 1$ if $q<0$.
■ Fix $q>0$, then the smallest phase eigenvalue $\lambda_{*}(k)$ is well defined, i.e.

$$
\vartheta_{*}(k):=\min \left\{\vartheta_{j}(k) \in[0, \pi): \text { where } \lambda_{j}(k)=r_{j}(k) e^{i \vartheta_{j}(k)}\right\} .
$$

## Inside-Outside Duality

## Inside-Outside Duality (KIRSCH, LECHLEITER, PETERS)

■ If $q>0$, and

$$
\lim _{k_{0}-\epsilon<k \nearrow k_{0}} \vartheta_{*}(k)=0
$$

and

$$
\lim _{k_{0}+\epsilon>k \searrow k_{0}} \vartheta_{*}(k)=0
$$

for small enough $\epsilon>0$. Then $k_{0}>0$ is a transmission eigenvalue.

For $q<0$ the above hold if the limits are $\pi$.

- The converse hods true for at least the first eigenvalue provided that the contrast $q$ is perturbation of a sufficiently large or small constant.


## TE and Non-desctructive Testing

For a given (unknown) anisotropic media $A$, we find an isotropic homogenous media $a_{0}$ that has the first transmission eigenvalue the same as the (measured) first transmission eigenvalue for the anisotropic media. Monotonicity properties gives that this $a_{0}$ is between $A_{\max }$ and $A_{\text {min }}$.

Numerical Example: We consider $D:=[-1,1] \times[-1,1]$ and fix $n=2$

| $A$ | $\tau_{1}$ | Predicted $a_{0}$ |
| :---: | :---: | :---: |
| diag(5.5,6.5) | 1.9657 | 5.95 |
| diad(5,7) | 1.9696 | 5.79 |
| diag(6,6.5) | 1.9591 | 6.24 |
| diag(6,7) | 1.9547 | 6.45 |

## TE and Non-desctructive Testing

$D:=[-1,1]^{2}, A=\operatorname{diag}(5,6), n=2, \operatorname{void} D_{0}:=B_{\epsilon}(0), A_{0}=I, n_{0}=1$
目 Harris-Cakoni-Sun (2014)-Inverse Problems


Figure 5. Graph of first transmission eigenvalue $k_{1}$ v.s. the size of a (large) circular void for $A=\operatorname{diag}(5,6)$ and $n=2$, and $D$ the unit circle and the square $[-1,1] \times[-1,1]$.

Table 4. First TEV for various void sizes computed by the FEM

| $\epsilon$ | 0.2 | 0.19 | 0.18 | 0.17 | 0.16 | 0.15 | 0.14 | 0.13 | 0.12 | 0.11 | 0.1 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Circle | 9.53 | 9.27 | 9.02 | 8.77 | 8.54 | 8.31 | 8.08 | 7.86 | 7.64 | 7.43 | 7.22 |
| Square | 7.76 | 7.57 | 7.39 | 7.21 | 7.04 | 6.87 | 6.70 | 6.53 | 6.37 | 6.21 | 6.05 |

## Spectral Analysis of Transmission Eigenvalue Problem

Where in the complex plane do transmission eigenvalues lie?
囯 Hitrik-Krupchyk-Ola-Paivarinta (2011) - Math.
Research Letters
Comprehensive spectral theory for transmission eigenvalue problem for isotropic media.

## Robbiano (2013) - Inverse Problems

Comprehensive spectral theory with Weyl asymptotic bounds for transmission eigenvalue problem for anisotropic media.

## Department of <br> Mathematics

# Spectral Theory of the Transmission Eigenvalue Problem 

 for Spherically Stratified Media
## Spherically Stratified Medium

The transmission eigenvalue problem for spherically stratified media is to find nontrivial $v, w \in L^{2}(D), v-w \in H_{0}^{2}(D)$ such that

$$
\begin{array}{clr}
\Delta v+k^{2} v=0 & \text { in } & B \\
\Delta w+k^{2} n w=0 & \text { in } & B \\
w=v & \text { on } & \partial B \\
\frac{\partial w}{\partial r}=\frac{\partial v}{\partial r} & \text { on } & \partial B
\end{array}
$$

where $B:=\{x:|x|<a\}$.

Transmission eigenvalues are non-scattering frequencies.
The far field operator is not injective and does not have dense range.

## Spherically Stratified Medium

Restricting to spherically stratified solutions, we make the ansatz

$$
v(r)=a_{0} \frac{\sin k r}{k r} \quad w(r)=b_{0} \frac{y(r)}{r}
$$

where $y(r)$ is the unique solution of the ODE

$$
\begin{aligned}
& y^{\prime \prime}+k^{2} n(r) y=0 \\
& y(0)=0, \quad y^{\prime}(0)=1
\end{aligned}
$$

Since $y(a)=a_{0} j_{0}(a), y^{\prime}(a)=a_{0} j_{0}^{\prime}(a)$ we have that transmission eigenvalues are solutions to

$$
d(k):=\operatorname{Det}\left|\begin{array}{cc}
y(a) & \frac{\sin k a}{k} \\
y^{\prime}(a) & \cos k a
\end{array}\right|=0 .
$$

## Spherically Stratified Medium

$d(k)$ is an entire function of $k$ that is real for real $k$ and is bounded on the real axis. Hence if $d(k)$ is not a constant then there exist a countably infinite set of transmission eigenvalues.

## Theorem (Aktosun-Gintides-Papanicolaou)

If $d(k) \equiv 0$ then $n(r) \equiv 1$.

We now assume that $n(r) \not \equiv 1$. Then from the asymptotic expression

$$
d(k)=\frac{1}{k a^{2}}\left[\frac{1}{[n(0) n(a)]^{1 / 4}} \sin (k \delta) \cos (k a)-\left[\frac{n(a)}{n(0)}\right]^{1 / 4} \cos (k \delta) \sin (k a)\right]+O\left(\frac{1}{k^{2}}\right)
$$

as $k \rightarrow \infty$, if

$$
\delta=\int_{0}^{a} \sqrt{n(\rho)} d \rho \neq a
$$

there exist an infinite number of positive transmission eigenvalues.

## Complex Transmission Eigenvalues

Example: Let $n(r)=n_{0}^{2}$ where $0<n_{0} \neq 1$ is a constant.

- When $n_{0}=\frac{2}{3}$ we have that

$$
d(k)=-\frac{1}{k} \sin ^{3}\left(\frac{k a}{2}\right)\left[3+2 \cos \left(\frac{2 k a}{3}\right)\right]
$$

$d(k)$ has an infinite set of real and complex zeros.

- For $n_{0}=\frac{1}{2}$ and we have that

$$
d(k)=-\frac{2}{k} \sin ^{3}\left(\frac{k a}{2}\right)
$$

$d(k)$ has an infinite set of real zeros and no complex zeros.
Note: There are always complex eigenvalues for $n$ constant if non-spherical eigenfunctions are included. (very recently proved by Colton-Leung)

## Entire Functions - Definitions

## Definition

Let $M(r)$ denote the maximum modulus of the entire function $f(z)$ on $|z|=r$. Then $f(z)$ is of order $\rho$ if

$$
\limsup _{r \rightarrow \infty} \frac{\log \log M(r)}{\log r}=\rho .
$$

Roughly $|f(z)| \leq A e^{\tau|z|^{\rho}}$

## Definition

The entire function $f(z)$ of order $\rho=1$ is called a function of exponential type $\tau$ if

$$
\limsup _{r \rightarrow \infty} \frac{\log M(r)}{r}=\tau
$$

## Spherically Stratified Medium

Now assume that $n(a)=1$ and $n^{\prime}(a)=0$.
$1 d(k)$ is an even entire function of $k$ of order (at most) one.
2 If $\int_{0}^{a} \rho^{2}[1-n(\rho)] d \rho \neq 0, d(k)$ has a zero of order two at $k=0$.
Thus, by the Hadamard factorization theorem, we have that

$$
d(k)=c k^{2} \prod_{j=1}^{\infty}\left(1-k^{2} / k_{j}^{2}\right)
$$

where $\left\{k_{j}\right\}$ are the zeros of $d(k)$ (including multiplicities) and $c$ is a constant. From

$$
d(k)=\frac{1}{k[n(0)]^{1 / 4}}\left\{\sin k\left(a-\int_{0}^{a} \sqrt{n(\rho)} d \rho\right)+O\left(\frac{1}{k}\right)\right\}
$$

as $k \rightarrow \infty$ along the positive real axis we have that $c[n(0)]^{1 / 4}$ is known. Hence, under the above assumptions, the transmission eigenvalues (real and complex!) determine $[n(0)]^{1 / 4} d(k)$.

## The Inverse Spectral Problem

As we have just seen, under appropriate assumptions the transmission eigenvalues determine $[n(0)]^{1 / 4} d(k)$. In order to determine $n(r)$ from $[n(0)]^{1 / 4} d(k)$ we need an integral representation of the solution to

$$
\begin{gathered}
y^{\prime \prime}+k^{2} n(r) y=0 \\
y(0)=0, \quad y^{\prime}(0)=1 .
\end{gathered}
$$

Using the Liouville transformation

$$
\begin{gathered}
\xi:=\int_{0}^{r} \sqrt{n(\rho)} d \rho \\
z(\xi):=[n(r)]^{1 / 4} y(r)
\end{gathered}
$$

We arrive at

$$
\begin{gathered}
z^{\prime \prime}+\left[k^{2}-p(\xi)\right] z=0 \\
z(0)=0, \quad z^{\prime}(0)=[n(0)]^{-1 / 4}
\end{gathered}
$$

where

$$
p(\xi):=\frac{n^{\prime \prime}(r)}{4[n(r)]^{2}}-\frac{5}{16} \frac{\left[n^{\prime}(r)\right]^{2}}{[n(r)]^{3}} .
$$

## The Inverse Spectral Problem

The solution of

$$
\begin{gathered}
z^{\prime \prime}+\left[k^{2}-p(\xi)\right] z=0 \\
z(0)=0, \quad z^{\prime}(0)=[n(0)]^{-1 / 4}
\end{gathered}
$$

can be represented in the form

$$
z(\xi)=\frac{1}{[n(0)]^{1 / 4}}\left[\frac{\sin k \xi}{k}+\int_{0}^{\xi} K(\xi, t) \frac{\sin k t}{k} d t\right]
$$

for $0 \leq \xi \leq \delta$ where $\delta=\int_{0}^{a} \sqrt{n(\rho)} d \rho$, and $K(\xi, t)$ is the unique solution of the Goursat problem

$$
\begin{aligned}
& K_{\xi \xi}-K_{t t}-p(\xi) K=0, \quad 0<t<\xi<\delta \\
& K(\xi, 0)=0, \quad 0 \leq \xi \leq \delta \\
& K(\xi, \xi)=\frac{1}{2} \int_{0}^{\xi} p(s) d s, \quad 0 \leq \xi \leq \delta
\end{aligned}
$$

A. KIRSCH (2011), An Introduction to the Mathematical Theory of Inverse Problems, Springer.

## The Inverse Spectral Problem

## Theorem (Rundell-Sacks)

Let $K(\xi, t)$ satisfy the above Goursat problem. Then $p \in C^{1}[0, \delta]$ is uniquely determined by the Cauchy data $K(\delta, t), K_{\xi}(\delta, t)$.

Now recall the determinant

$$
d(k):=\operatorname{Det}\left|\begin{array}{cc}
y(a) & -\frac{\sin k a}{k} \\
y^{\prime}(a) & -\cos k a
\end{array}\right|=0 .
$$

From the Liouville transformation and the representation for $z(\xi)$ we have that

$$
\begin{gathered}
y(a)=\frac{1}{[n(0)]^{1 / 4}}\left[\frac{\sin k \delta}{k}+\int_{0}^{\delta} K(\delta, t) \frac{\sin k t}{k} d t\right] \\
y^{\prime}(a)=\frac{1}{[n(0)]^{1 / 4}}\left[\cos k \delta+\frac{\sin k \delta}{2 k} \int_{0}^{\delta} p(s) d s+\int_{0}^{\delta} K_{\xi}(\delta, t) \frac{\sin k t}{k} d t\right]
\end{gathered}
$$

## The Inverse Spectral Problem

Note that the asymptotic formulas for $d(k)$ gives us $\delta$. The above formula now gives us

$$
\begin{equation*}
\frac{\ell \pi}{a} d\left(\frac{\ell \pi}{a}\right)=\frac{(-1)^{\ell+1}}{[n(0)]^{1 / 4}}\left[\sin \frac{\ell \pi \delta}{a}+\int_{0}^{\delta} K(\delta, t) \sin \frac{\ell \pi t}{a} d t\right] \tag{1}
\end{equation*}
$$

and

$$
\begin{align*}
\frac{\ell \pi}{a} d\left(\frac{\ell \pi}{\delta}\right) & =-y(a) \frac{\ell \pi}{\delta} \cos \frac{\ell \pi a}{\delta} \\
& +\frac{\sin \frac{\ell \pi a}{\delta}}{[n(0)]^{1 / 4}}\left[(-1)^{\ell}+\frac{\delta}{\ell \pi} \int_{0}^{\delta} K_{\xi}(\delta, t) \sin \frac{\ell \pi t}{\delta} d t\right] \tag{2}
\end{align*}
$$

## The Inverse Spectral Problem

- Since $\left\{\sin \frac{\ell \pi t}{a}\right\}$ is complete in $L^{2}[0, \delta]$ if $\delta \leq$ a we have from (1) that $K(\delta, t)$ (and hence $y(a)$ ) is known.
- From (2) and the completeness of $\sin \frac{\ell \pi t}{\delta}$ in $L^{2}[0, \delta]$ we have that $K_{\xi}(\delta, t)$ is known.

The Rundell-Sacks Theorem now implies that $p(\xi)$ is uniquely determined for $0 \leq \xi \leq \delta$ from a knowledge of $[n(0)]^{1 / 4} d(k)$.
From this we can now easily determine $n(r)$.

## The Inverse Spectral Problem

## Theorem (Colton-Leung)

Assume that $n \in C^{3}[0, a], n(a)=1$ and $n^{\prime}(a)=0$. If $0<n(r)<1$ for $0<r<a$ the transmission eigenvalues (including multiplicity) with spherically symmetric eigenfunctions, uniquely determine $n(r)$.

## Theorem (Cakoni-Colton-Gintides)

Assume that $n \in C^{1}[0, \infty), 0<n(r)<1$ or $n(r)>1$, and that $n(0)$ is known. All the transmission eigenvalues uniquely determine $n(r)$.

The only extension of the above theorem to the case of more general domains $D$ is for $n$ constant. More specifically, $n$ is uniquely determined from a knowledge of the smallest positive transmission eigenvalue provided it is known a priori that either $n>1$ or $0<n<1$.

## Complex Transmission Eigenvalues Again

The previous result on the inverse spectral problem requires that $n(a)=1$ and $n^{\prime}(a)=0$. However, our previous example on the existence of complex transmission eigenvalues was for $n$ constant, i.e. having a jump across the boundary.

Recall that if $\delta=\int_{0}^{a} \sqrt{n(\rho)} d \rho \neq$ a real transmission eigenvalues always exist.

We now examine the existence of complex transmission eigenvalues when $n(a)=1$ and $n^{\prime}(a)=0$.

## Complex Transmission Eigenvalues Again

Let $n_{+}(r)$ denote the number of zeros of an entire function $f(z)$ in the right half plane with $|z|<r$.

## Theorem (Cartwright-Levinson)

Let the entire function $f(z)$ of exponential type be such that

$$
\int_{-\infty}^{\infty} \frac{\log ^{+}|f(x)|}{1+x^{2}} d x<\infty
$$

and suppose that

$$
\limsup _{y \rightarrow \pm \infty} \frac{|f(i y)|}{|y|}=\tau .
$$

Then

$$
\lim _{r \rightarrow \infty} \frac{n_{+}(r)}{r}=\frac{\tau}{\pi}
$$

## Complex Transmission Eigenvalues Again

## Definition

The number $\tau / \pi$ is called the density of zeros in the right half plane.
We now again consider

$$
y^{\prime \prime}+k^{2} n(r) y=0, \quad y(0)=0, \quad y^{\prime}(0)=1
$$

and use the Liouville transformation

$$
\xi:=\int_{0}^{r} \sqrt{n(\rho)} d \rho, \quad z(\xi):=[n(r)]^{1 / 4} y(r)
$$

As previously, we have the representation

$$
z(\xi)=\frac{1}{[n(0)]^{1 / 4}}\left[\frac{\sin k \xi}{k}+\int_{0}^{\xi} K(\xi, t) \frac{\sin k t}{k} d t\right]
$$

and again define

$$
d(k):=\operatorname{Det}\left|\begin{array}{cc}
y(a) & -\frac{\sin k a}{k} \\
y^{\prime}(a) & -\cos k a
\end{array}\right|
$$

## Complex Transmission Eigenvalues Again

Integrating by parts in the expression for $z(\xi)$ now yields

$$
\begin{aligned}
& d(k)=\frac{-1}{k[n(0)]^{1 / 4} n(a)^{1 / 4}}\left[\sin ((\delta-a) k)-\frac{K(\delta, \delta)}{k} \cos ((\delta-a) k)\right. \\
& \left.\quad+\frac{K_{\tau}(\delta, \delta)-K_{\xi}(\delta, \delta)}{2 k^{2}} \sin ((\delta-a) k)+\frac{n^{\prime \prime}(a)}{8 k^{2}} \sin ((\delta+a) k)+O\left(\frac{1}{k^{3}}\right)\right]
\end{aligned}
$$

where again $\quad \delta:=\int_{0}^{a} \sqrt{n(\rho)} d \rho$.
Thus $d(k)$ is of type $(\delta+a)$ and the leading term $\sin ((\delta-a) k)$ generates an infinite set of positive real zeros with density $|\delta-a| / \pi$. However, if $n^{\prime \prime}(a) \neq 0$, from the Cartwright-Levinson theorem the density of all zeros in the right half plane is $(\delta+a) / \pi$.

## Theorem (Colton-Leung-Meng)

Suppose that $n \in C^{2}[0, a]$ with $n(a)=1$ and $n^{\prime}(a)=0$ and $\delta \neq 1$. Then, under the extra assumption that $n^{\prime \prime}(a) \neq 0$, there exist infinitely many real and infinitely many complex transmission eigenvalues.


[^0]:    Q Cakoni - Haddar (2012) - Transmission Eigenvalues, Inside Out, MSRI.

